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Adaptive controlled superconducting magnetic energy storage devices for performance enhancement of wind energy systems



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ABSTRACT

This research paper introduces the Generalized Continuous Mixed P-Norm Sub-Band Adaptive Filtering (GCMPSAF) algorithm, designed for efficient online control of Superconducting Magnetic Energy Storage Devices (SMESDs) in Wind Energy Systems (WESs). The primary objective of this algorithm is for minimizing power ripples in WESs. The Wind Energy System (WES) under consideration is tied to the IEEE 39 bus system, with the Superconducting Magnetic Energy Storage Device (SMESD) integrated at the point of common coupling. The GCMPSAF algorithm is applied to update or adapt proportional-integral (PI) controller gains of SMESD interface circuits. The proposed algorithm is an enhanced version of the CMPN by adding the sub-band filtering algorithm effect. It depends mainly on the actuating error signal, and it has a variable step size of the CMPN. The detailed modeling of the whole system is presented, including measured wind speed data, detailed switching techniques, a drive train model of the turbine, and real SMESD. The efficacy of the proposed SMESD has been validated through a comparative analysis with the Least Mean Square-SMESD approach, under conditions of varying and unpredictable wind speeds. The simulation results produced by the PSCAD software are used to evaluate the study's validity. The utilization of controlled SMESDs has the potential to significantly enhance the power quality of WESs.

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1. Introduction

1.1. Research problem

The renewable energy systems (RESs) have garnered significant interest worldwide in recent times, owing to their potential to offer sustainable and clean sources of energy. There are several reasons beyond the significant potential of the RESs, such as an surge in

fuel prices, its reduction possibility, the trend toward sustainable development goals, environmental issues, and some political concerns [1]. A broad consensus is existed in promoting fast and effective decarbonization, as it was an extensive object of the Sharm El-Sheikh climate change conference (COP27). Therefore, electricity generation from RESs should push forward to satisfy economic decarbonization [2]. Wind energy conversion systems are considered the second rank of largest RESs after global hydropower. The annual global wind energy report states that new installations of 94 GW are added, and the total wind energy capacity reaches 837 GW in 2021. Based on these statistical analyses, interconnection of wind systems with power grids has reached a state of high saturation in recent times leading to arising of many problems, including power systems reliability, dynamics, stability, operation, and control. Power quality issue is considered one of the challenges face wind systems. The mechanical power of wind systems is based on the wind speed cube, and this wind speed possesses an inter-

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Abbreviations

| | | | |
|----------|---|--------|---|
| ANN | Artificial neural network | PLL | Phase-locked loop |
| CMPN | Continuous Mixed P-Norm | PI | Proportional-integral |
| ESDs | Energy storage devices | PCC | Point of common coupling |
| FS | Fixed-speed | RESs | Renewable energy systems |
| FLCs | Fuzzy logic controllers | SMESDs | Superconducting Magnetic Energy Storage Devices |
| GCMPNSAF | Generalized Continuous Mixed P-Norm Sub-Band Adaptive Filtering | VS | Variable-speed |
| IGBT | Insulated gate bipolar transistor | VSC | Voltage source converter |
| LMS | Least mean square | WES | Wind Energy System |
| | | WT | Wind turbine |

mittent and sophisticated nature profile. This is easily resulting in more output power fluctuations in wind energy systems. Generally, wind energy systems (WESs) may contain fixed-speed (FS) or variable-speed (VS) wind turbines (WTs) or both. However, FSWTs suffer from the most significant mechanical stress of their shaft, leading to a higher fluctuation in their output power generation [3]. This affects more variation of reactive power of such generators. Moreover, these power fluctuations have an adverse impact on the voltage response at the point of common coupling (PCC) of the WESs. This leads to a deep reduction in the power quality concern of such WESs. The primary aim is to achieve an efficient control of energy storage devices, with the aim of mitigating output power fluctuations and enhancing power quality of WESs.

1.2. Related works

The primary goal of many research projects has been to reduce real power ripples of WESs, enhancing the power quality concern of such renewable energy sources. It is claimed that a blade pitch angle control can help decrease such power ripples of WESs [4]. The mechanical nature of this control mechanism, however, results in a delayed response time. Additionally, if it reacts quickly to reduce these ripples, the blades will be under much compressive stress. The maximum power of WESs is tracked, and its fluctuations are decreased using the rotational inertia methodology [5]. Nevertheless, when a significant disruption occurs, it struggles with delayed transient behavior. Various RESs, including waves and wind energy conversion systems, are used to reduce active-power fluctuations during the islanded mode of operation [6]. Moreover, the ramp control approach is employed in the heat ventilation air conditioning system to regulate the active power of WESs [7]. There are many distributed generation sources like wind and photovoltaic generators cooperating together with the purpose of enhancing the active power profile [8]. Also, the energy of rotating masses of permanent magnet synchronous wind generator can be utilized in power ripples minimization of WESs [9], but the scientific model of the scheme does not consider uncertainty effects. In [10], a fuzzy logic control scheme with the help of a low pass filter to enhance power responses. Moreover, a classical approach is implemented to reduce the power ripples of WESs declining system disturbances, but it depend on on a linear model for the system [11]. In [12], The model predictive controller has been employed in order to effectively mitigate output power fluctuations in WESs, and it depends on a quadratic programming approach. However, like all conventional optimization methods, it relies on initial conditions, solver accuracy, and may stuck into local minima point. A two-layer control scheme was recently proposed to optimize power balance between generation and load demand, while also enhancing the wind power profile. However, the procedure involved in this approach is complex and time-

consuming [13]. The hydrogen system is used to stabilize wind power fluctuations [14].

The integration of storage devices with WESs has a high influence on the overall effectiveness of active power smoothing techniques. Battery energy storage devices (ESDs) are extensively used to minimize power fluctuations of WESs. In [15], comprehensive reviews and surveys of various control techniques of batteries are demonstrated to minimize power ripples of such systems. Effective control of the battery system is crucial for optimizing the system cost. The dual batteries ESDs are implemented to smooth WESs power fluctuations with better economic benefits [16]. In [17], a wavelet analysis method is proposed for battery ESDs to investigate different components of power ripples. In addition, battery ESDs can be sized based on the model predictive control scheme to minimize power ripples of WESs [18]. A deep reinforcement learning methodology is proposed to control wind energy systems with battery ESDs to achieve the same target [19]. In [20], the impact of battery ESDs on power ripple minimization of WESs is investigated by considering the initial value of the state of charge, switching frequency, and high pass filters. Although the battery ESD is a more cost-effective device, it has numerous drawbacks, including its delayed response time due to chemical processes, the constraints on voltage and current, and its negative environmental impact. Despite the high cost of super-capacitor systems, these are widely used to minimize power ripples and enhance the dynamic responses of wind energy systems [21–23]. In addition, flywheel ESDs are considered a candidate solution to smooth power ripples of WESs. In this regard, flywheel ESDs have various advantages, including reduced cost, greater power capacity, more reliability, true transient performance, and longer lifespans [24–26]. Despite the merits of flywheel ESDs, these suffer from high friction and windage losses leading to a lower efficiency [3]. On the other hand, superconducting magnetic ESDs (SMESDs) represent adequate storage devices for power ripples minimization of WESs. SMESDs possess several merits, like high efficiency [27], quick transient behavior [28], and unlimited cycles of charging/discharging processes [29]. The cost of superconducting material plays an important role in this industry, and there are many research funds oriented toward decreasing the material cost [30,31]. Various controllers have been implemented to control SMESDs. To enhance the efficiency of photovoltaic power plants, SMESDs are controlled by fuzzy logic controllers (FLCs) [32]. Moreover, FLCs are utilized to control SMESDs with the purpose of improving the microgrids stability [29]. Although FLCs work with the nonlinearity of the system and no need to a system model, they rely on fuzzy rules and gains that are appropriately designed. In addition, an artificial neural network (ANN) control strategy is applied to SMESDs to enhance the stability of WESs [33,34]. Despite the merits of ANN controllers that deal with the system's nonlinearity, they establish complex structures and require a long time to perform training and testing processes. Proportional-Integral (PI) controllers are widely utilized in various applications

in modern times. This is a result of its broad stability range and low-cost applications. The systems nonlinearity, unpredictability, and susceptibility to parameter modification, however, are drawbacks. In order to obtain optimal gains for PI controllers, especially for complex, nonlinear systems where transfer functions or state space models are complex to obtain, various metaheuristic optimization algorithms are employed for their design [35–39]. Although these approaches are useful and excellent tools for designing PI controllers, they have several drawbacks, such as a complicated methodology, a heavy memory requirement, and a lengthy optimization phase. Adaptive PI controllers are strong choices for online tuning of controller gains, as a consequence, optimization procedures are not required. This helps in cutting down on the time and effort needed for the lengthy design process. Several adaptation techniques, including adaptive weight algorithm [40], affine projection [41], and variable mixing norm [42], were employed to update PI controller gains. The tremendous development of adaptive filtering algorithms, including a quick and accurate response, represents the principal incentive of the authors to utilize a generalized continuous mixed p-norm sub-band adaptive filtering (GCOMPNSAF) algorithm to fully control SMESDs for minimizing power ripples of WESs.

1.3. Main contribution

Numerous technical issues, such as those involving communications and signals, have been successfully solved by the adaptation methods and associated algorithms [43,44]. Adaptive Filtering Algorithms (AFAs) typically rely on an error signal, which represents the difference between a reference signal and the actual output signal. Most AFAs depend on the concept of least mean square (LMS) error [45]. In addition, some approaches rely on the error signal's p-norm [46]. The utilization of AFAs has been widely adopted to improve the dynamic performance of Renewable Energy Systems [47,48]. It is essential to find the right balance between an algorithm's accuracy and complexity. Therefore, in this article, an intelligent GCOMPNSAF algorithm is presented. The proposed algorithm is an improved version of the CMPN algorithm, achieved through the incorporation of sub-band filtering techniques. It depends mainly on the actuating error signal, and it has a variable step size of the CMPN [49]. The GCOMPNSAF algorithm possesses various advantages over other AFAs, such as individual weight factor SAF, normalized logarithmic SAF, and improved proportionate normalized SAF [49]. Some benefits of the GCOMPNSAF include quick convergence, a straightforward process, and less complicated computing. Therefore, the GCOMPNSAF algorithm is implemented in this study to automatically adapt PI controllers gains of SMESDs with the determination of flattening the output power of WESs.

The current study involves the connection of a WES to the IEEE 39 bus New England system, with a SMESD tied to the WES at the point of common coupling (PCC). A VSC is employed within the SMESD to effectively control both PCC and DC-link voltages, with an additional DC chopper circuit utilized to regulate the active power of the SMESD. The proposed adaptive PI controllers are responsible for controlling the converters. In order to achieve realistic results, actual wind speed data obtained from Hokkaido Island is used for dynamic simulations. The WT model is thoroughly considered and actual SMESD is utilized, with simulation analyses performed through the use of PSCAD software [50]. The validity of the controlled SMESD is established by conducting a comparative analysis of its results with those obtained using other methods, particularly under the nonlinearity conditions of the system under study. The active power output from the WES can be made more uniform and their ripple can be significantly reduced with the adoption of the suggested control approach, enhancing the power quality of WESs.

The paper organization is done as follows: System modeling is demonstrated in Section 2. The WES model is introduced in Section 3. A SMESD control strategy is demonstrated in Section 4. The GCOMPNSAF algorithm is fully detailed in Section 5. The findings are illustrated in Section 6, with a focus on research. Finally, Section 7 brings study conclusions.

2. The system model

The power network chosen for this investigation is the IEEE 39 bus standard system. It illustrates the New England power network in a small form. Utilizing such massive systems is done primarily to provide a realistic evaluation of the WES. The graphical representation of such system is shown in Fig. 1 [3]. It contains 39 buses and there are 19 load buses. Bus 31 is known as the slack-bus and the system has 10 generators. Active power for the load and generation is 6098.1 and 6140.81 MW, respectively. As a choice for load modeling, constant current and admittance is used. The generators' active power output varies from 250 to 1000 MW. A 10 MW WES is tied to bus 30 as well as the synchronous generator number 10, which provides 240 MW. The proposed controlled SMESD is connected to the WES PCC. The statistics for generators, buses, and loads are shown in Tables 1-3, respectively [3].

3. The WES modelling

The following mathematical formulas can be used to express the mechanical power obtained from wind turbines (WTs) [33]:

$$P_M = 0.5\rho\pi R^2 V_w^3 C_p(\lambda, \beta) \quad (1)$$

where P_M stands for mechanical power of a WT, ρ is the air density, R is the radius of a WT, V_w represents the wind velocity. Power coefficient (C_p) is expressed by tip-speed ratio λ and blade angle β as follows:

$$C_p = 0.5(\lambda_i - 0.022\beta^2 - 5.6)e^{-0.17\lambda_i} \quad (2)$$

$$\lambda = \frac{\omega_B R}{V_w}, \lambda_i = \frac{3600 R}{1609 \lambda} \quad (3)$$

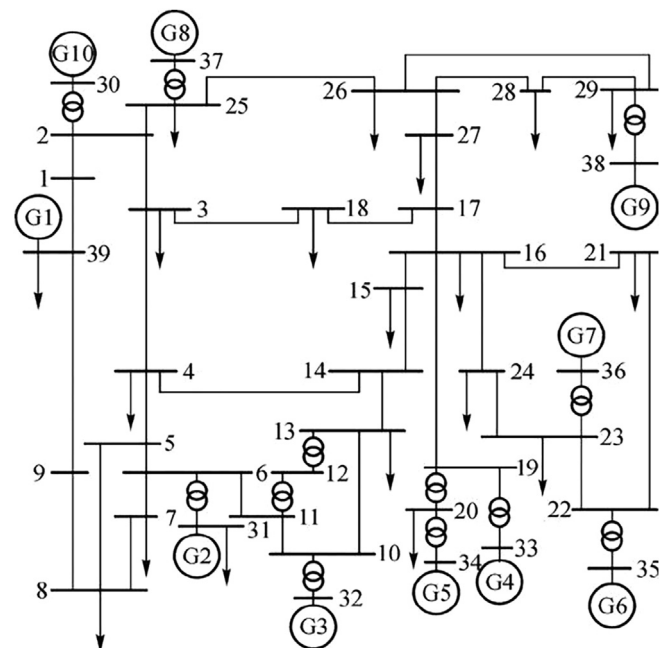


Fig. 1. IEEE 39 bus system.

Table 1
Generators Data.

| Bus | Total apparent Power (MVA) | Generated real power (MW) | Bus voltage(p.u) |
|-----|----------------------------|---------------------------|------------------|
| 30 | 289.590 | 250 | 1.04750 |
| 31 | 557.270 | 520.810 | 0.9820 |
| 32 | 681.600 | 650 | 0.98310 |
| 33 | 641.490 | 632 | 0.99720 |
| 34 | 534.360 | 508 | 1.01230 |
| 35 | 683.830 | 650 | 1.04930 |
| 36 | 569.070 | 560 | 1.06350 |
| 37 | 540 | 540 | 1.02780 |
| 38 | 830.310 | 830 | 1.02650 |
| 39 | 1003 | 1000 | 1.030 |

In (3), ω_B stands for an angular velocity of a WT. The three-drive train model is implemented to represent WTs and it provides a high level of accuracy and has a significant impact on transient responses. Its detailed characteristics are demonstrated in Table 4. Inertia constants of a wind turbine, gear box and generator are represented by H_{WT} , H_{GB} , and H_G . Parameters K_{HGB} and K_{CGB} stands for shaft spring constant between hub-gear box and gear box-generator, respectively. D_{WT} , D_{GB} , and D_G devotes self-damping coefficient of such components. d_{HGB} and d_{CGB} represents mutual damping between them. The WES consists of five squirrel cage induction generators having a rated power of 2 MVA each.

Table 2
Buses Data.

| Bus | Bus name | Base kV | Voltage magnitude (p.u) | Voltage angle (degree) |
|-----|----------|---------|-------------------------|------------------------|
| 1 | 1 | 345 | 1.04500 | -8.5400 |
| 2 | 2 | 345 | 1.04100 | -5.8200 |
| 3 | 3 | 345 | 1.02500 | -8.6800 |
| 4 | 4 | 345 | 1.00100 | -9.6800 |
| 5 | 5 | 345 | 1.00300 | -8.6700 |
| 6 | 6 | 345 | 1.00600 | -8.0100 |
| 7 | 7 | 345 | 0.99500 | -10.1900 |
| 8 | 8 | 345 | 0.99400 | -10.6900 |
| 9 | 9 | 345 | 1.02800 | -10.4100 |
| 10 | 10 | 345 | 1.01500 | -5.4800 |
| 11 | 11 | 345 | 1.01100 | -6.3400 |
| 12 | 12 | 345 | 0.99800 | -6.3100 |
| 13 | 13 | 345 | 1.01200 | -6.1600 |
| 14 | 14 | 345 | 1.00900 | -7.7200 |
| 15 | 15 | 345 | 1.01300 | -7.7900 |
| 16 | 16 | 345 | 1.02900 | -6.2300 |
| 17 | 17 | 345 | 1.0300 | -7.3400 |
| 18 | 18 | 345 | 1.02700 | -8.2800 |
| 19 | 19 | 345 | 1.04900 | -1.0600 |
| 20 | 20 | 345 | 0.99100 | -2.0500 |
| 21 | 21 | 345 | 1.0300 | -3.8200 |
| 22 | 22 | 345 | 1.04900 | 0.6400 |
| 23 | 23 | 345 | 1.04400 | 0.4400 |
| 24 | 24 | 345 | 1.03500 | -6.1100 |
| 25 | 25 | 345 | 1.04400 | -4.1800 |
| 26 | 26 | 345 | 1.04500 | -5.4700 |
| 27 | 27 | 345 | 1.03200 | -7.4900 |
| 28 | 28 | 345 | 1.04700 | -1.9300 |
| 29 | 29 | 345 | 1.04800 | 0.8400 |
| 30 | Gen 10 | 22 | 1.04800 | -3.3900 |
| 31 | Gen 2 | 22 | 0.98200 | 0.0 |
| 32 | Gen 3 | 22 | 0.98300 | 2.5100 |
| 33 | Gen 4 | 22 | 0.99700 | 4.1600 |
| 34 | Gen 5 | 22 | 1.01200 | 3.1400 |
| 35 | Gen 6 | 22 | 1.04900 | 5.5900 |
| 36 | Gen 7 | 22 | 1.06400 | 8.300 |
| 37 | Gen 8 | 22 | 1.02800 | 2.5100 |
| 38 | Gen 9 | 22 | 1.02700 | 7.9100 |
| 39 | Gen 1 | 345 | 1.0300 | -10.1500 |

4. SMESD and its control circuits

As an energy storage device in power or renewable energy systems, SMESD has beneficial properties including high efficiency, rapid reaction, no degradation during repeated operation, and more. Then, a wide range of SMESD applications are anticipated. Nevertheless, it includes multiple issues like reliability verification and lowering costs for effective SMESD utilization. There are numerous initiatives worldwide to develop SMESDs. In this paper, a practical SMESD is connected to the WES PCC. Nominal power and energy storage of the SMESD is 10 MW and 20 MJ, respectively. It was successfully installed at Hosoo power plant, Japan [51]. It consists of a three-phase transformer, VSC, DC link capacitor, a DC chopper circuits and superconducting coil. The DC link capacitor has a capacitance of 15 mF. The SMESD coil has a self-inductance L_s of 21.1H. The SMESD components are demonstrated in Fig. 2. Stored energy E_s of SMESD and its nominal power P_s is written by the following formulas:

$$E_s = 0.5L_s I_s^2 \tag{4}$$

$$P_s = \frac{dE_s}{dt} = L_s I_s \frac{dI_s}{dt} = V_s I_s \tag{5}$$

where I_s and V_s are the coil instantaneous current and voltage.

4.1. Voltage source converter

In Fig. 2, the voltage source converter (VSC) is presented. It is a three-phase converter with six IGBTs. The VSC control is based on a cascaded control scheme, as shown in Fig. 3. There are two adaptive PI controllers located in outer loops to regulate V_{PCC} and V_{DC} . In addition, there are two adaptive PI controllers located in inner loops to regulate direct and quadrature (dq) axis currents. The V_{PCC} is measured to feed a phase-locked loop (PLL) system, which extracts an angle θ_s . This angle is essential for the abc-dq0 frames conversion. To create the firing pulses of IGBTs, $V_{a,b,c-n}$ are compared with a triangle signal with a 1.2 kHz frequency.

Table 3
Loads Data.

| Bus No. | Load real power (MW) | Load reactive power (Mvar) |
|---------|----------------------|----------------------------|
| 3 | 322 | 2.40 0 |
| 4 | 500 | 184 |
| 7 | 233.80 | 84 |
| 8 | 522 | 176 |
| 12 | 8.5 | 88 |
| 15 | 320 | 153 |
| 16 | 329 | 32.30 0 |
| 18 | 158 | 30 |
| 20 | 628 | 103 |
| 21 | 274 | 115 |
| 23 | 247.50 | 84.600 |
| 24 | 308.6 0 | -92 |
| 25 | 224 | 47.20 0 |
| 26 | 139 | 17 |
| 27 | 281 | 75.500 |
| 28 | 206 | 27.60 0 |
| 29 | 283.50 | 26.90 0 |
| 31 | 9.2 0 | 4.60 0 |
| 39 | 1104 | 250 |

Table 4
WTs Characteristics.

| H_G (s) | 0.14190 | D_{WT} (pu) | 0.0220 |
|----------------|---------|----------------|--------|
| H_{GB} (s) | 0.08060 | D_{CB} (pu) | 0.0220 |
| H_{WT} (s) | 1.92770 | D_G (pu) | 0.010 |
| K_{HGB} (pu) | 54.750 | d_{HGB} (pu) | 3.50 |
| K_{CGB} (pu) | 1834.10 | d_{CGB} (pu) | 10 |

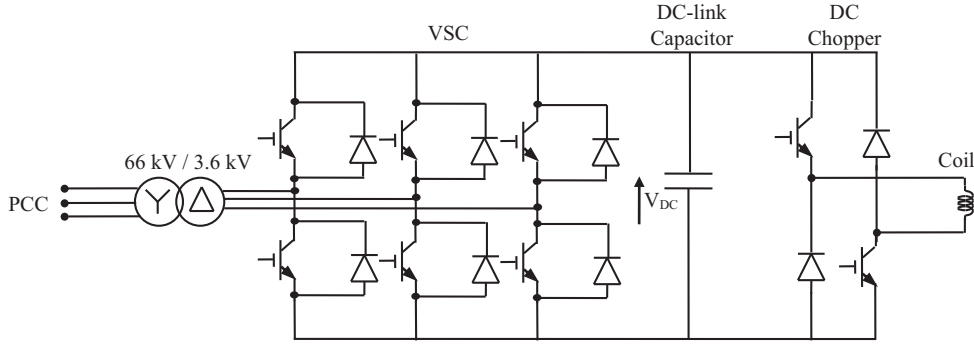


Fig. 2. SMESD components.

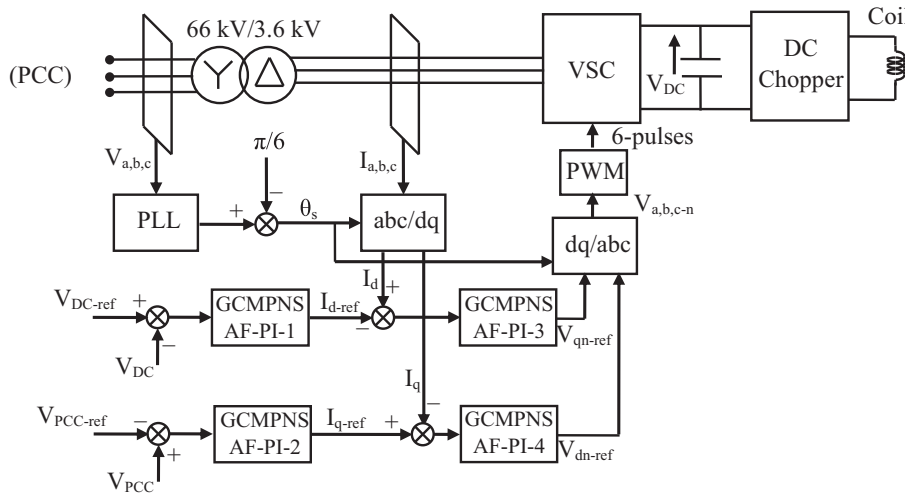


Fig. 3. VSC control scheme.

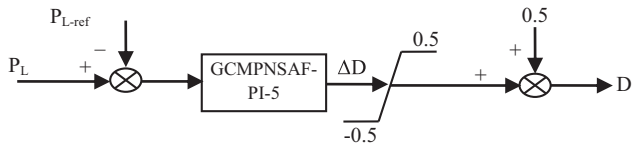


Fig. 4. DC chopper circuit control.

4.2. DC chopper circuit

A two-quadrant DC-DC converter is used in this SMESD to control the active power at the PCC (P_L). A control strategy depends on adjusting chopper duty cycle (D), where it lies in the range $[0, 1]$ to carry out chopper cycles of SMESD. Hence, one adaptive PI controller is used to regulate P_L to its reference value P_{L-ref} . Fig. 4 points out the DC chopper circuit control. Lastly, the duty cycle signal enters a PWM control circuit to generate delaying pulses of IGBTs [33].

5. AFAs

5.1. GCM PNSAF algorithm

Mixed-norm adaptive filters are a class of AFAs that come in a number of variations. There is some combination between AFAs like least mean square (LMS) with least mean fourth to produce

a LM mixed-norm or LMMN algorithm [52]. Also, a LMS is combined with a least absolute deviation to form the robust MN algorithm [53]. Hence, the robust normalization MN algorithm is presented [54] and it relies on error norm minimization approach by:

$$C(k) = L(k)E\{er^2(k)\} + (1 - L(k))E\{|er(k)|\} \quad (6)$$

where $C(k)$ stands for the objective function, k represents an iteration index, $L(k)$ represents parameter varying from 0 to 1. $er(k)$ represents an error signal of the AFA by:

$$er(k) = d(k) - y(k) \quad (7)$$

$$y(k) = x^T(k)we(k) \quad (8)$$

where $y(k)$ stands for the AFA output vector, $x(k)$ represents the input vector, $d(k)$ represents the reference signal vector, and $we(k)$ represents a present weight vector of AFA [45]. In [55], a p-norm strategy is applied to LMS and the standard norm is 2. The CMPNSAF is further developed and is modeled as follows [46]:

$$C(k) = \int_1^2 L_k(p) E\{|er(k)|^p\} dp \quad (9)$$

where $L_k(p)$ devotes to the probability function and its constraints are expressed as follows:

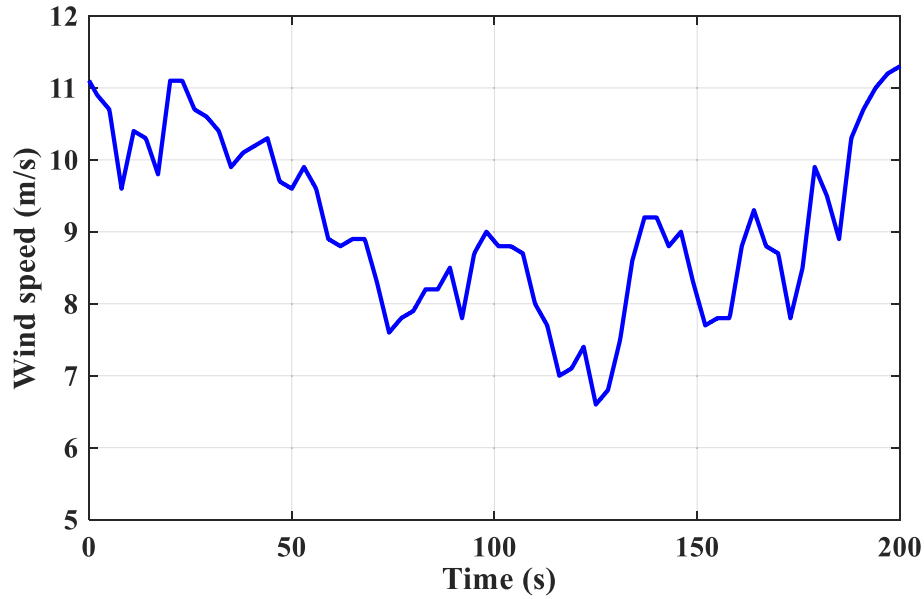


Fig. 5. Wind speed profile.

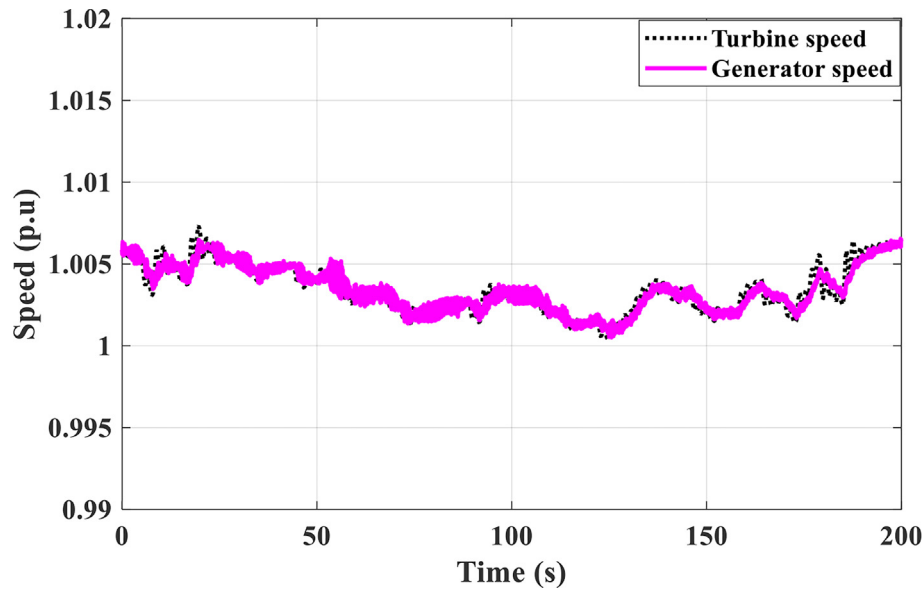


Fig. 6. Wind turbine and generator speed profile.

$$\int_1^2 L_k(p) dp = 1 \tag{10}$$

Weight vector of CMPNAFA can be updated by the following formula:

$$we(k+1) = we(k) - \mu \nabla_{we(k)} C(k) \tag{11}$$

where μ represents a step size. $\nabla_{we(k)} C(k)$ is mathematically expressed as follows:

$$\frac{\partial C(k)}{\partial we_i(k)} = \int_1^2 L_k(p) \frac{\partial}{\partial we_i(k)} E\{|er(k)|^p\} dp \tag{12}$$

The expectation $E\{|er(k)|^p\}$ can be approximately equal to $|er(k)|^p$. Then, the weight vector of CMPNAFA can be developed by:

$$we(k+1) = we(k) + \mu \gamma_k \text{sign}(er(k))x(k) \tag{13}$$

$$\gamma_k = \int_1^2 p L_k(p) |er(k)|^{p-1} dp \tag{14}$$

where γ_k represents a variable-step size and can be written as follows:

$$\gamma_k = \frac{(2|e(k)| - 1) \ln(|e(k)|) - |e(k)| + 1}{(\ln(e(k)))^2} \tag{15}$$

Therefore, in this article, an intelligent GCMPSAF algorithm is presented. The proposed algorithm is an enhanced version of the CMPN by adding the sub-band filtering algorithm effect. It depends mainly on the actuating error signal and it has a variable step size of the CMPN [49]. The GCMPSAF algorithm possesses various advantages over other AFAs such as individual weight factor SAF,

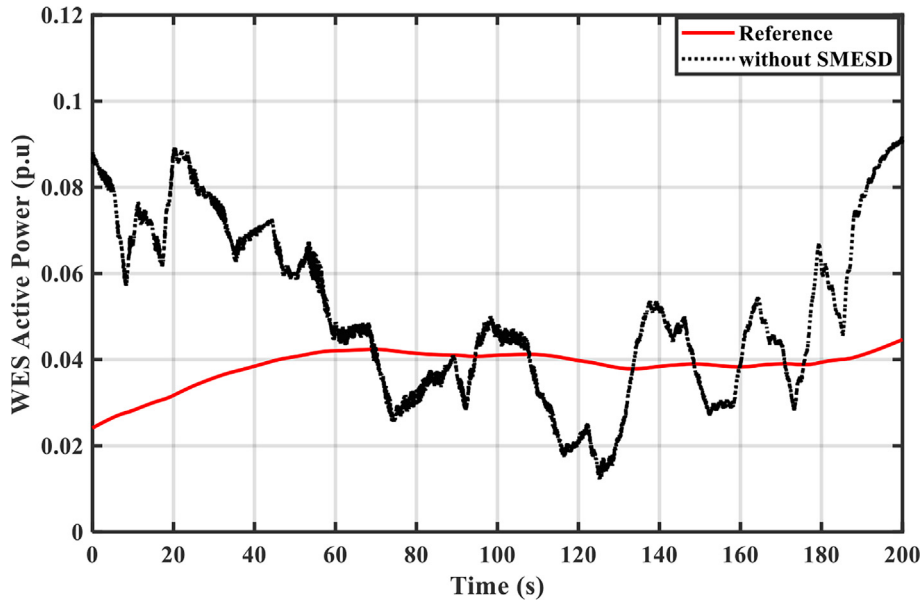


Fig. 7. WES active power.

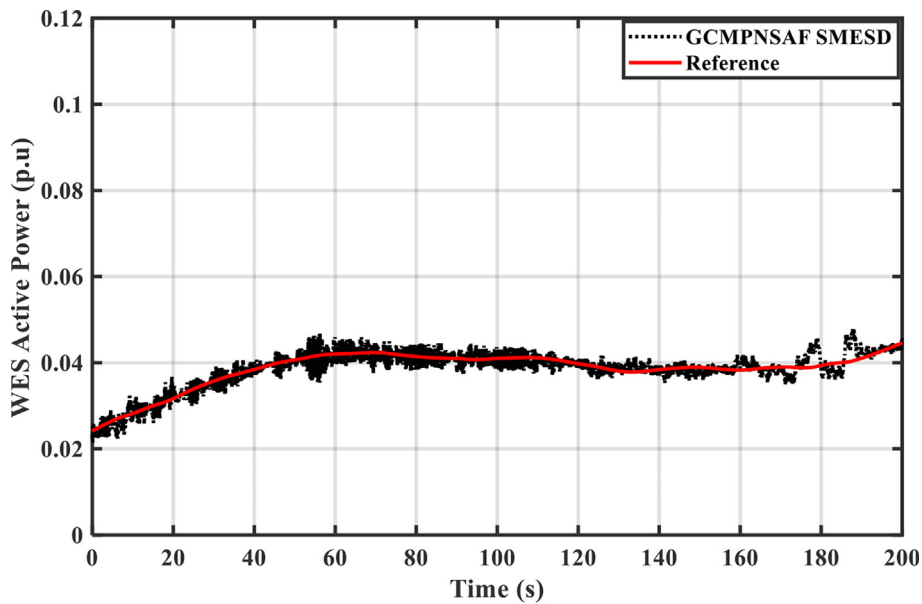


Fig. 8. WES active power using the proposed GCMPSAF controlled SMESD.

normalized logarithmic SAF, and improved proportionate normalized SAF [49]. Some benefits of the GCMPSAF include quick convergence, a straightforward process, and less complicated computing. The weight vector of the GCMPSAF algorithm can be written as follows:

$$w_e(k + 1) = w_e(k) + \frac{\mu \gamma_k \text{sign}(e(k))x(k)}{\sqrt{x^T(k)x(k) + \gamma_k}} \quad (16)$$

In this article, the controllers gains of the SMESD are adapted instantaneously by (16). $k_p(k)$ and $k_i(k)$ is adapted as follows:

$$k_p(k + 1) = k_p(k) + \Delta k_p(k) \quad (17)$$

$$k_i(k + 1) = k_i(k) + \Delta k_i(k) \quad (18)$$

$$\Delta k_p(k) = \Delta k_i(k) = \frac{\mu \gamma_k \text{sign}(e(k))x(k)}{\sqrt{x^T(k)x(k) + \gamma_k}} \quad (19)$$

where $e(k)$ is an input signal to the adaptive PI controller and $x(k)$ stands for a real signal. The initial conditions and controllers parameters are chosen to maintain the system stability.

5.2. LMS algorithm

One of the fundamental AFAs is the traditional LMS method. It has a relatively straightforward structure and uses less processing steps [3]. These characteristics, however, have an impact on performance and do not guarantee getting superior outcomes. It is based on an error signal to adapt the weight vector of the AFA and can be written by:

$$we(k + 1) = we(k) + \mu er(k)x(k) \tag{20}$$

where μ is equal to 0.012. Controllers gains are updated by:

$$\Delta k_p(k) = \Delta k_i(k) = \mu er(k)x(k) \tag{21}$$

In this paper, a SMESD controlled by adaptive LMS-PI controllers are effectively compared with the proposed adaptive GCMPSAF-PI controllers under the WES variability and uncertainty.

6. Simulation results and discussion

In this study, a 10 MW WES is tied to bus 30 as well as the synchronous generator number 10, which provides 240 MW. The proposed controlled SMESD is connected to the WES PCC. The detailed modeling is applied to all components, including wind turbines, wind generator, and power converters. The detailed modeling of electronic switches is considered. The simulation analyses are conducted using PSCAD software, which is well-suited for conducting transient analyses in power system studies. Simulation time inter-

val and simulation run is chosen 10 μ s and 200 s, respectively. To ensure a realistic study, the WES model incorporates measured wind speed data obtained from Hokkaido Island, as illustrated in Fig. 5, where it has a wide range variation from 6.6 to 11.3 m/s. Fig. 6 demonstrates the wind turbine and generator speeds profile. It can be realized that the generator speed exceeds the synchronous speed and lies in a satisfactory range. The WES desired active power (P_{L-ref}) and the active power without using SMESD are shown in Fig. 7. P_{L-ref} is determined by following the generator power on a low pass filter. The WES active power using adaptive SMESD based on the GCMPSAF and the LMS algorithms are illustrated in Figs. 8 and 9, respectively. It is noteworthy that the proposed technology achieves better damping of active power ripples in WES with minimum fluctuations compared to the LMS technology. Such power ripple is minimized by 15% at specific times using the proposed technology. The SMESD active power is demonstrated in Fig. 10, which indicates the dynamic interaction between the SMESD and the grid through the charging and discharging stages of the storage device. To evaluate the power quality, the

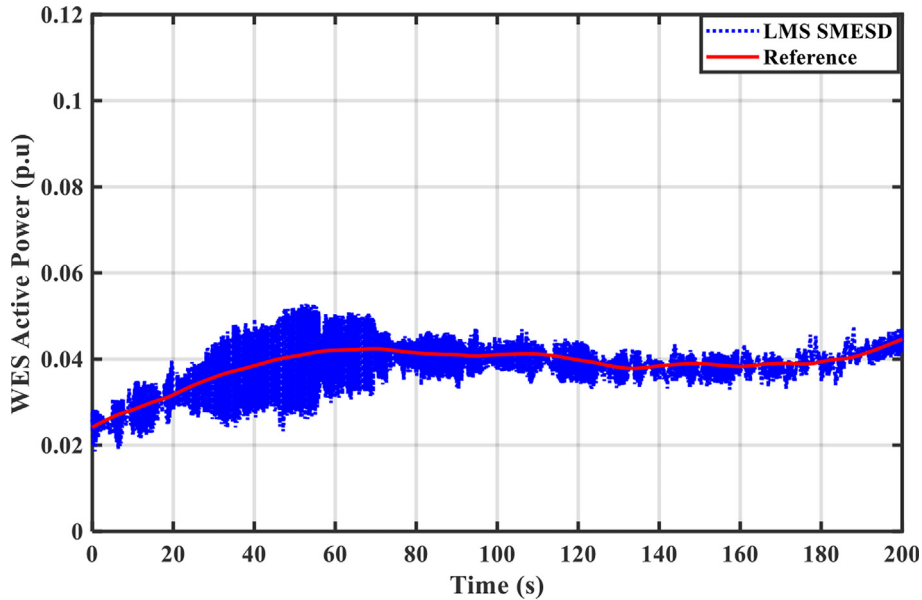


Fig. 9. WES active power using the LMS controlled SMESD.

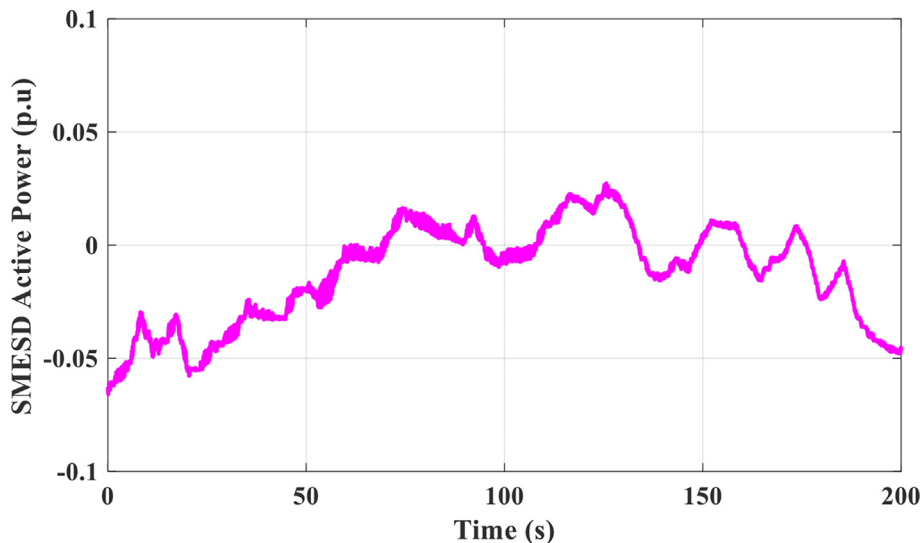


Fig. 10. SMESD active power.

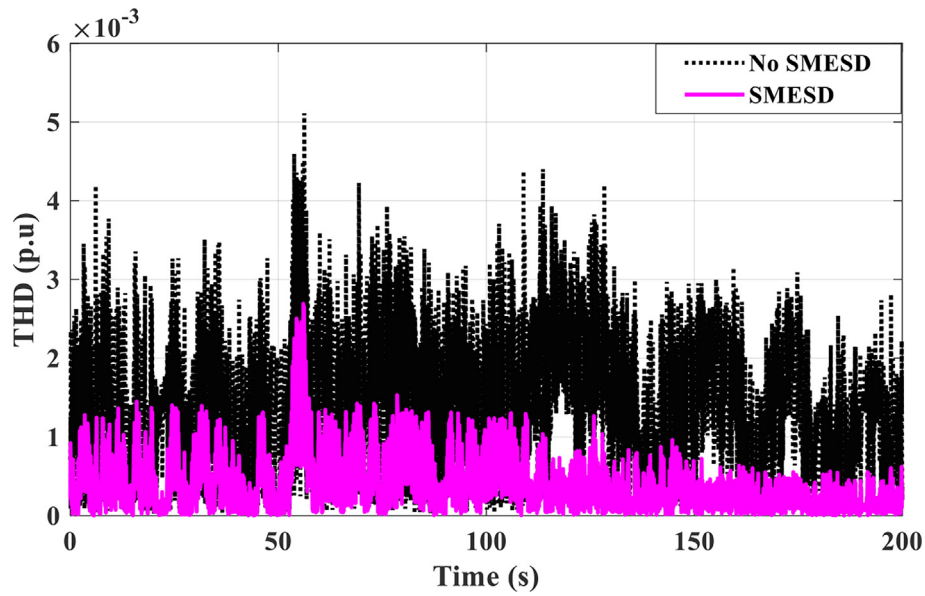


Fig. 11. THD of WES active power.

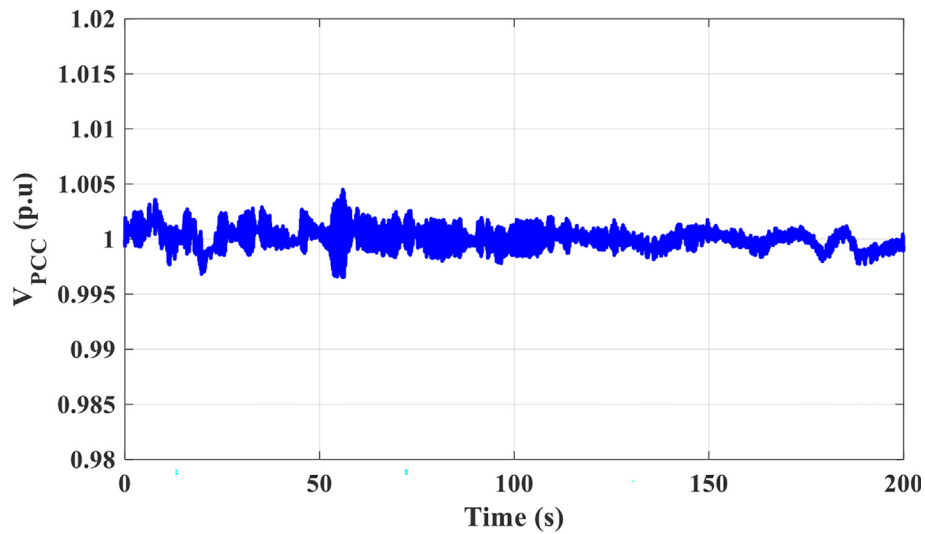


Fig. 12. V_{PCC} response.

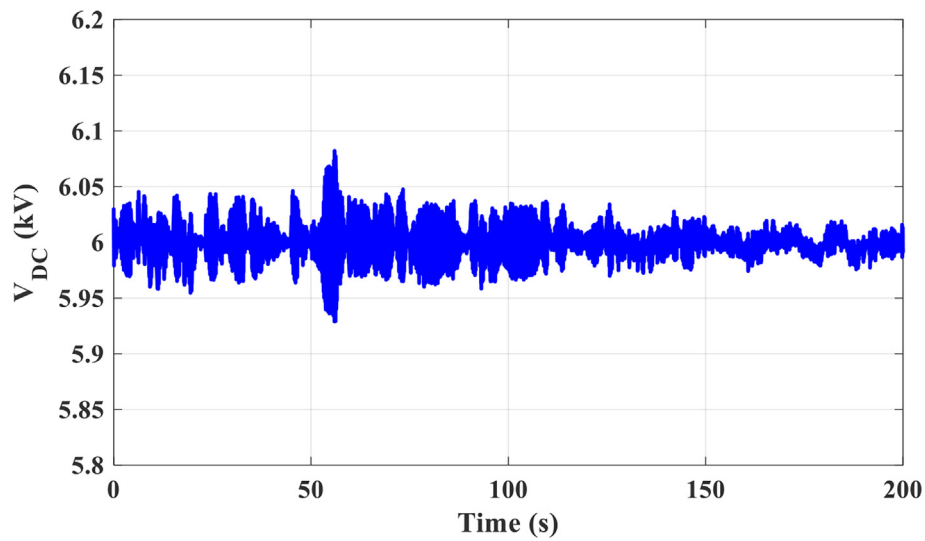


Fig. 13. V_{DC} response.

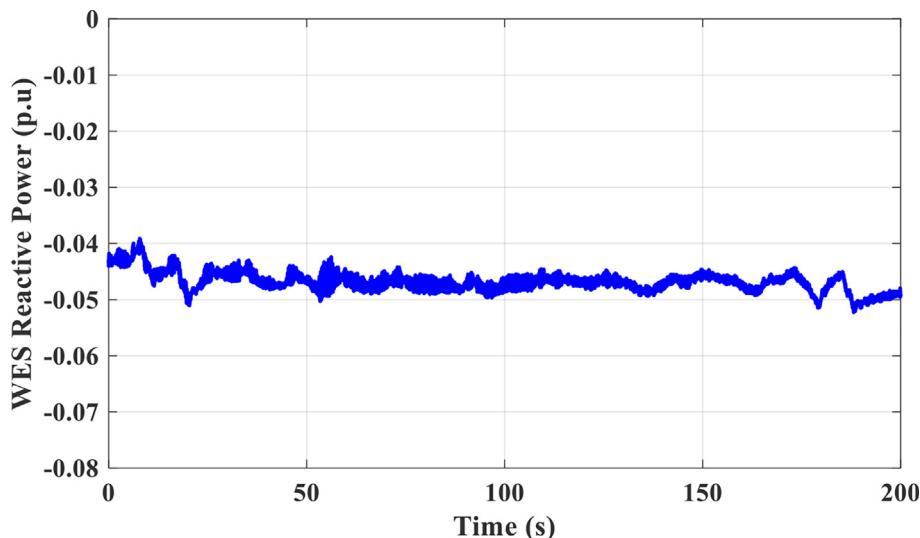


Fig. 14. WES reactive power response.

total harmonic distortion (THD) factor is analyzed for the active power response. Fig. 11 depicts the THD value of active power with and without SMESD. Notably, its value is significantly lower than that achieved without SMESD. This leads to a power quality improvement of the WESs. The V_{PCC} response is effectively adjusted to the nominal value using the controlled SMESD under wind speed uncertainty and variability, as illustrated in Fig. 12. Moreover, Fig. 13 shows the V_{DC} response of SMESD and it is kept at its rated value. The WES reactive power is plotted in Fig. 14 and it can be noted that this reactive power is supported from the SMESD. Therefore, the implementation of adaptive controlled SMESDs shall smooth the WES active power and enhance its power quality issue.

7. Conclusion

This paper has presented a novel GCMPSAF algorithm to fully online control SMESDs for minimizing power ripples of WESs. The proposed WES is connected to the IEEE 39 bus system and the SMESD is tied to the point of common coupling. The GCMPSAF algorithm is applied to adapt all PI controller gains of SMESD interface circuits. The proposed algorithm is an enhanced version of the CMPN by adding the sub-band filtering algorithm effect. It depends mainly on the actuating error signal and it has a variable step size of the CMPN. The detailed modeling of the whole system is presented, including measured wind speed data, detailed switching techniques, drive train model of turbine, and real SMESD. The simulation results have verified outperform of the proposed SMESD technology over that obtained using the LMS-SMESD technology under the system nonlinearity and uncertainty. The power ripple of the WES is minimized by 15% at specific times using the proposed technology. The proposed control scheme is fully adaptive and there is no need to any optimization methods leading to saving of time and efforts. The proposed technology can help in smoothing the WES active power and enhancing the system power quality. It shall be applied to several microgrids and smart grid applications. Moreover, the developments of adaptive algorithms will be extensively applied to power systems control and dynamics.

Declaration of Competing Interest

The authors declare that they have no known competing financial interests or personal relationships that could have appeared to influence the work reported in this paper.

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